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(54) Speech intelligibility enhancement system and method.

(57) To enhance the intelligibility of speech, the consonant sounds are intensified and, in effect, their intensity equalised to that of the vowel sounds in a speech waveform. A short-time estimate of the relative spectral shape of an input speech signal is determined by envelope detectors (24) operating on the outputs of band pass filters (20). Control means are provided to respond to such relative spectral shape estimate by dynamically controlling a modification of the spectral shape of the actual speech signal so as to produce a modified output speech signal, the control means comprising a combination matrix (28) operating on the outputs of the envelope detectors (24) with a matrix of coefficients and producing weighted signals (29) as control signals. The control signals (29) act on gain selecting logic (30) to determine the gains of multipliers (31) through which respective different portions of the frequency spectrum of the input speech are coupled to a summation circuit (32) producing the consonant - enhanced output speech signal, the respective different portions of the frequency spectrum of the input speech being produced by a bank of filters (20) supplying the envelope detectors (24) or by a set of different filters (26).

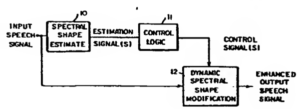

FIG. 1

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SPEECH INTELLIGIBILITY ENHANCEMENT SYSTEM AND METHOD

This invention relates generally to the enhancement of the intelligibility of speech and more particularly to the enhancement of the consonant sounds of speech.

It is desirable in many applications to enchance the intelligibility of speech when the speech has been processed electronically as, for example, in hearing aids, public address systems, radio or telephone communications, and the like. Although it is helpful to enhance the presentation of both vowel and consonant sounds, generally it appears that, since the intelligibility characteristics of speech depend to such a significant extent on consonant sounds, it is primarily desirable to enhance the intelligibility of such consonants.

Several approaches have characterised recent research into such intelligibility problems, particularly with respect to the hearing aid field. One approach has been to take the high frequency sounds in speech and transpose them to lower frequencies so that they fall within the band of normal hearing acuity, leaving the low frequency sounds unprocessed. Such approaches are discussed, for example, in the article "A Critical Review of Work on Speech Analysing Hearing Aids" by A. Risberg, IEEE Trans. Audio and Electroacoustics, Vol. AU-17. No. 4, December 1969, pp. 290-297. The degree of success of such an approach appears to be quite limited and overall improvement in perceiving consonants, for example, was relatively small.

An alternative approach, akin to the frequency lowering technique, has been to slow down the overall speech, i.e., to lower the frequencies of the overall speech waveform thereby presenting the higher frequency content at lower frequencies within the listener's normal hearing band. If such a technique is used in

real time, segments of the speech have to be removed in order to make room for the remaining temporally expanded segments and such process can generate distortion in the speech. Such techniques are discussed in the article "Moderate Frequency Compression for the Moderately Hearing Impaired", M. Mazor et al., J. Acoust. Soc. Am., Vol. 62, No. 5, November 1977, pp. 1273-1278. Although some slight improvement has been observed using such frequency compression techniques for up to about 20% frequency compression, for example, it was also noted that a further increase in frequency compression only tended to reduce intelligibility.

A basic problem with both high frequency transposition techniques and frequency compression schemes is that they tend to distort the temporal-frequency patterns of speech. Such distortion interferes with the cues needed by the listener to perceive the speech features. As a result such approaches tend to meet with only limited success in enhancing speech intelligibility.

Another approach to speech intelligibility enhancement is one which preserves the bandwidth of the speech and, instead, modifies the level and dynamic range of the speech waveform. The goal of such a speech processing approach is to make full use of the listener's high frequency hearing abilities. The hearing abilities of the hearing impaired are described, for example, in the article, "Differences in Loudness Response of the Normal and Hard of Hearing Ear at Intensity Levels Slightly above Threshold", by S. Reger, Ann. Otol., Rhinol., and Laryngol., Vol. 45, 1936, pp. 1029-1036. In this study of hearing impairment it was noted that soft sounds could not be perceived because of the loss in sensitivity, but that more intense sounds were perceived as having near-normal loudness. This phenomenon, sometimes referred to as "recruitment", has formed a motivation for improved hearing aid designs. Thus, an approach that

tends to preserve the speech bandwidth and improves intelligibility by modifying the speech waveform dynamics and spectral energy appears to be a more effective approach than frequency transposition or frequency compression techniques because the features of the speech are better preserved. Although such an approach has achieved some success, as reported in the article "Signal Processing to Improve Speech Intelligibility for the Hearing Impaired" by E. Villchur, J. Acoust. Soc. Am., Vol. 53, pp. 1646-1657, June 1973, improvement is still needed to provide the most effective enhancement of the intelligibility of speech, particularly in the enhancement of consonant sounds.

The system of the invention provides an improved and effective enhancement of the reproduction of consonant sounds by emphasising the spectral content of consonants so as to intensify the consonant sound and, in effect, to equalise its intensity with that of vowel sounds, the latter sounds tending to achieve a normal intensity much greater than the normal consonant intensity. In accordance with the broadest approach of the invention, the system thereof processes an input speech signal by determining a short-time estimate of the spectral shape. The term "spectral shape" as used herein is intended to mean the spectral content of the input speech signal as a function of frequency relative to the spectral content at a specified frequency, or a specified frequency region, of the input speech signal. The term "spectral content" is intended to mean, for example, the energy content of the signal as a function of frequency, the envelope of the signal at a plurality of frequencies or in a plurality of frequency bands, the short-time Fourier transform coefficients of the signal, and the like. Control means are provided in response to such relative spectral shape estimate for dynamically controlling a modification of the spectral shape of the actual speech signal so

as to produce an output speech signal.

Such modification can be achieved, for example, by first estimating the short-time spectral shape of the overall frequency spectrum of the input speech signal. One way of providing such estimate, for example, is to determine the spectral contents of different selected frequency bands within the overall spectrum, (e.g., the energy in each band, the envelope in each band, the Fourier transform coefficients in each band, or the like) relative to the spectral content of one or more reference bands. This determination can be achieved by using Fourier transform techniques, filtering techniques, and the like. The estimated spectral shape of the overall input speech signal spectrum, however achieved, is then used to control, or modify, the spectral shape of the actual input signal, as for example, by modifying the spectral content of one or more frequency bands of the input signal (which may or may not coincide with the previously mentioned selected frequency bands) to produce the output speech signal. The term "short-time" spectral shape, as used herein, means the spectral shape over a selected short time interval of between about 1 millisecond to about 30 milliseconds.

The invention will now be described in more detail, solely by way of example with reference to the accompanying drawings wherein:-

FIG. 1 is a broad block diagram of a system embodying the invention;

FIG. 2 is a more specific block diagram of a system embodying the invention;

FIG. 3 is a further, more specific block diagram of a system embodying the invention;

FIG. 4 is a specific block diagram of an alternative embodiment of the invention which is a modification of that depicted in FIG. 3;

FIG. 5 is a still more specific block diagram of a system embodying the invention;

FIG. 6 illustrates schematically and more specifically a combination matrix circuit of the embodiment of FIG. 5;

FIG. 7 is a more specific block diagram of an embodiment of the invention;

FIG. 8 is a further specific block diagram of another alternative embodiment of the invention; and

FIG. 9 is a graphical representation of the amplitude envelope characteristics as a function of time as obtained at the exemplary point in the embodiment of the invention depicted in FIG. 8.

FIG. 1 depicts a broad block diagram of a system for processing an input signal in accordance with the techniques of the invention. As can be seen therein, an input speech signal is supplied to means 10 for estimating the spectral shape of the input speech signal. Such spectral shape estimation, when determined, provides one or more estimation signals for supply to a suitable control logic means 11 which is responsive to such spectral shape estimate for suitably controlling the dynamic modification of the spectral shape of the actual input speech signal via appropriate spectral shape modification means 12 to produce an enhanced output speech signal, as desired. The output speech can then be appropriately used wherever desired. For example, the output speech signal may be supplied to a suitable transmitter device or a system, e.g., a public address system or voice communication system, a radio broadcast transmitter, etc., or to a suitable receiver device, e.g., a hearing aid, a telephone receiver, an earphone, a radio, etc.

A particular approach in accordance with the general approach shown in FIG. 1 is depicted in FIG. 2 wherein the speech signal is supplied to a bank of filters 20, i.e., a plurality of bandpass filters for

providing a plurality of frequency bands within the overall speech frequency spectrum of the input speech signal. An estimate of the spectral content in each frequency band relative to the spectral content in one or more reference bands is made in spectral shape estimation means 21 for supplying a plurality of estimation signals to control means 22 which in turn supplies one or more control signals for dynamically modifying the overall spectral shape of the input speech signal. For example, the control signal may select one of a plurality of different filters for modifying the spectral content of the input speech signal, the selection thereof depending on the particular estimate that was made. Alternatively, for example, a plurality of control signals may be generated to control a plurality of separate filters each of which corresponds to a selected pass band of the frequency spectrum of the input speech signal. The pass bands of the filter bank used to modify the actual input speech signal may or may not correspond to the pass bands of the filter bank so used to form the spectral shape estimates.

FIG. 3 depicts a more specific block diagram of the above approach wherein the input speech signal is supplied to a selected number N of bandpass filters 20, designated as BP_1 through BP_N. The spectral shape of the input speech signal is determined by detecting the envelope characteristics of the outputs of each of the bandpass filters 20 using suitable envelope detectors 24. A control logic unit 22 is responsive to the outputs of envelope detectors 24 and provides a control signal which is used to select one suitable enhancement filter from a plurality of M such filters 25, identified as filters F_1 to F_M, each having selected characteristics for dynamically modifying the shape

of the overall spectrum of the input speech signal which is supplied thereto. The output from a selected one of such enhancement filters 25 thereby provides a desired consonant enhanced output speech signal.

Alternatively, FIG. 4 depicts a system similar to that of FIG. 3 wherein the selection control logic 22 provides a plurality of control signals, each supplied to one of a plurality of N band-pass filters 26, indentified as BP'_1 to BP'_N, for modifying the spectral characteristics of the input speech signal in each pass-band. The modified outputs from each filter 26 are appropriately summed at a summation circuit 27 to provide the desired consonant enhanced output speech signal.

A specific embodiment of the speech enhancement of FIG. 3 is depicted in FIG. 5 wherein envelope detectors 24 produce a plurality of envelope detector signals $X_1 \ldots X_N$ which are supplied to combination matrix logic 28 to produce weighted signals $W_1 \ldots W_N$ each of which represents the ratios 29 as depicted. One stage of the combination logic matrix 28 for producing the weight W_1 is shown more specifically in FIG. 6 wherein a plurality of preselected constant coefficients $a_{11} \ldots a_{NN}$ and $b_{11} \ldots b_{NN}$ are used to multiply the envelope detected signals $X_1 \ldots X_N$. The summation of the multiplier outputs corresponding to the "a" coefficients is divided by the summation of the multiplier outputs corresponding to the "b" coefficients to form the weight W_1, as shown. Similar matrix steps are used to form weights $W_2 \ldots W_N$. The weights $W_1 \ldots W_N$ are supplied to selection circuitry for selecting an appropriate filter 25 in accordance therewith.

In a specific exemplary embodiment of the invention depicted in FIGS. 3 and 5, three band-pass filters 20 were chosen so that BP_1 covered 2-4 KHz, BP_2

covered 1-2 kHz, and BP_3 covered 0.5-1 kHz. The combination matrix 28 was chosen to give weights $W_1=X_1/X_3$, $W_2=X_2/X_3$, and $W_3=1$. In such case, for example, the weights are determined by a comparison of the relative energies among the bands, e.g., the envelope detected signal from one of the filters (e.g., X_3) is used as a reference and the energies in the other bands (e.g., X_1 and X_2) are, in effect, compared with such reference to provide the desired weights. For example, when the energy in a particular band (X_1) is large compared with that in the reference band (X_3), the weight W_1 is greater than unity, when the energies are equal the weight is unity, and when the energy is less than the reference band energy the weight is less than unity. For the specific weights discussed in the above example the coefficient matrices are as follows:-

$$\begin{bmatrix} a_{11} & a_{12} & a_{13} \\ a_{21} & a_{22} & a_{23} \\ a_{31} & a_{32} & a_{33} \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$\begin{bmatrix} b_{11} & b_{12} & b_{13} \\ b_{21} & b_{22} & b_{23} \\ b_{31} & b_{32} & b_{33} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 \\ 0 & 0 & 1 \\ 0 & 0 & 1 \end{bmatrix}$$

The enchancement filter selection circuit at the output was chosen to contain three filters, one being a high-pass filter emphasising the region above 2.5 kHz, one being a band-pass filter emphasising the region from 1 kHz to 2.5 kHz, and the third being an all-pass filter having unity gain at all frequencies.

The weights were then used by the selection circuit to form a composite filter which had a gain of 1 below 0.5 kHz and which gave a 3:1 dynamic range expansion when the associated weight for a given frequency band was above a pre-selected threshold. This composite filter was updated every millisecond to give the dynamic spectral shape modification desired. In a similar manner, FIG. 7 shows a more specific embodiment of the approach depicted in FIG. 4 wherein the input speech signal, as in the embodiment of FIG. 5, is supplied to band-pass filters 20 and envelope detectors 24. Combination matrix logic 28 combines the envelope detected outputs X_1, X_2 ... X_N, in a selected manner, as discussed above, to produce a plurality of weighting signals W_1 . . . W_N in the same general manner as discussed above with respect to FIGS. 5 and 6. In this case the weighting factors W_1 . . . W_N are used to select suitable gain constants G_1 . . . G_N at gain select logic 30 for multiplying the filtered outputs of bandpass filters 26, designated as BP'_1 . . . BP'_N, as in FIG. 4, which filters separate the input speech signal into selected spectral bands. The filtered outputs from bandpass filters 26 are multiplied by the corresponding gains G_1 . . . G_N at multipliers 31, the outputs of which are added at summation circuit 32 to produce the consonant enhanced output speech signal.

The bandwidths of the input signals to multipliers 31 need not necessarily coincide with the bandwidths of the input signals to envelope detectors 24 and in the general case shown in FIG. 7 different portions of the frequency spectrum may be used for each bank of filters 20 and 26. In a simplified version thereof, the pass bands may coincide in which case the outputs of bandpass filters 20 can be supplied directly to

multipliers 31 (as well as to envelope detectors 24) and the filter bank 26 eliminated.

In the embodiment of FIG. 7 the coefficients $a_{11} \ldots a_{NN}$ and $b_{11} \ldots b_{NN}$ are selected empirically and the weights are then used to provide gains which produce independent dynamic range expansions in the selected frequency bands. One effective approach is to select the gain by comparing the weight W_i with a preselected threshold and to provide for unity gain when the weight is below the threshold and to provide an increased gain at or above such threshold. The increased gain may be selected logarithmically, i.e., in accordance with a selected power of the weight involved. For example, for suitable expansion on a db (logarithmic) scale the gain can be selected in accordance with the second power, i.e., W_i^2 when above the selected threshold, although effective expansion may also be achieved ranging from the first power (W_i) to the third power (W_i^3).

While the pass bands of the filters used in the above described embodiments of FIGS. 2-7 may be selected to provide pass bands which are clearly separated one from another, the degree of separation does not appear to significantly affect the consonant enhancement, although excessive separation would appear to have disadvantages in some application. Further, some degree of overlapping of the pass bands does not appear to have an adverse effect on the overall enhancement operation.

In a specific example of the invention depicted in FIG. 7, for example, four band pass filters 20 are used (filters 26 were eliminated) such that BP_1 covers 2-5 kHz, BP_2 covers 1-2 kHz, BP_3 covers 0.5-1 kHz and BP_4 covers 0-0.5 kHz. The coefficients "a" and "b" are selected so as to provide weights $W_1=X_1/X_3$, $W_2=X_2/X_3$, $W_3=1$ and $W_4=1$. In each case the envelope detected outputs of each band relative to the

envelope detected output of a reference band determines the weight. Thus, the weights W_1, W_2 and W_3 are determined by the envelope detected outputs X_1, X_2 and X_3 relative to the envelope detected output X_3, while W_4 is determined by the envelope detected output X_4 relative to X_4. Accordingly, the coefficients are selected as follows:

$$\begin{bmatrix} a_{11} & a_{12} & a_{13} & a_{14} \\ a_{21} & a_{22} & a_{23} & a_{24} \\ a_{31} & a_{32} & a_{33} & a_{34} \\ a_{41} & a_{42} & a_{43} & a_{44} \end{bmatrix} \quad \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

$$\begin{bmatrix} b_{11} & b_{12} & b_{13} & b_{14} \\ b_{21} & b_{22} & b_{23} & b_{24} \\ b_{31} & b_{32} & b_{33} & b_{34} \\ b_{41} & b_{42} & b_{43} & b_{44} \end{bmatrix} \quad \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

The gains are selected as follows:

$$\text{If } W_1<2, \quad G_1= 1$$
$$W_1\geq 2, \quad G_1= W_1{}^2/_4$$

$$\text{If } W_2<2, \quad G_2= 1$$
$$W_2\geq 2, \quad G_2= W_2{}^2/4$$

$$G_3=G_4=1 \quad \text{(always)}$$

A further improvement can be made in the approach of the invention by using the modifications discussed with reference to FIGS. 8 and 9 which are designed to take into better account the background noise present

in the input speech signal. If an estimate of such background noise is made and the effects of such noise is appropriately removed in the spectral shape estimate control operation the consonant enhancement can be further improved.

A technique for such operation is depicted in FIG. 8 wherein the outputs of each of the bandpass filters 20 are supplied both to peak detectors 35 and to valley detectors 36. The peak detectors follow the peaks of the signal by rising rapidly as the signal increases but falling slowly when the signal level decreases. The valley detectors follow the minima of the signal by falling rapidly as the signal decreases but rising slowly when the signal level increases. The time constant of the peak detector decay is in general much shorter than that of the valley detector rise. Thus, the output waveforms from such detectors tend to be of the exemplary forms shown in FIG. 9 wherein the sold line 37 represents an input to the detectors 35 and 36 from a bandpass filter 20, the dotted line 38 represents the peak detector output waveform and the dahsed line 39 represents the valley detector output waveform.

The valley detected output signal tends to represent the background noise present in the input speech signal and if such signal is subtracted at subtractors 40 from the peak detected output (which, in effect, represents the desired signal plus background noise), the signals $X_1 \ldots X_N$ provide improved spectral shape estimates which can then be suitably combined as in the combination matrix means 28 for providing the weighted signals $W_1 \ldots W_N$ as before.

While the specific implementations discussed above are disclosed to show particular embodiments of the invention, the invention is not limited thereto.

Modifications thereto within the spirit and scope of the invention will occur to those in the art. For example, instead of using discrete filters, as shown by the filter bands discussed above, other techniques for determining the spectral content in selected frequency bands can be used, such as fast Fourier transform (FFT) techniques, chirp-z (CZT) techniques and the like. Moreover, the spectral content need not be the envelope detected output but can be an energising detected output, the Fourier transform coefficients in a Fourier transform process, or other characteristics representative of the spectral content involved. Hence, the invention is not to be construed as limited to the particular embodiments described except as defined by the appended claims.

CLAIMS

1. A system for processing an input speech signal, characterised by

means (10) responsive to said input speech signal for estimating the short-time spectral shape of said input speech signal as a function of time;

control means (11) responsive to said shape estimate for providing one or more control signals; and

means (12) responsive to said one or more control signals for dynamically modifying the spectral shape of said input speech signal to produce an output speech signal.

2. A system in accordance with claim 1, characterised in that the estimating means (10) estimates the spectral content in each of a plurality of selected frequency bands relative to the spectral content in one or more of said frequency bands.

3. A system in accordance with claims 1 or 2, characterised in that said estimating means (10) includes

means (20) for separating said input speech signal into a plurality of selected frequency bands; and

means (21) responsive to the portions of said input speech signal in each of said frequency bands for estimating the spectral content in each of said frequency bands relative to the spectral content in a selected one or more of said frequency bands;

said control means (23) being responsive to the spectral content estimates in said frequency bands for producing said one or more control signals.

4. A system in accordance with claim 3, characterised in that said separating means is a bank of filters (20).

5. A system in accordance with claim 3, characterised in that said estimating means (21) includes

a plurality of envelope detection means (24) for detecting the envelope characteristics of said input speech signal in each of said frequency bands; and

said control means (22) is responsive to said envelope characteristics for providing said one or more control signals.

6. A system in accordance with claim 5, characterised in that said control means includes

means (28,29) responsive to said envelope characteristics for providing a plurality of weighting signals; and

means (22) responsive to said weighting signals for producing said one or more control signals.

7. A system in accordance with any preceding claim, characterised in that said modifying means includes

a plurality of filter circuits (25) each having a different characteristic over the frequency spectrum of said input speech signal; and

means (22) responsive to said one or more control signals for selecting one of said plurality of filter circuits (25) to modify said input speech signal so as to produce said output speech signal.

8. A system in accordance with any one of claims 2 to 6, characterised in that said modifying means includes

means (26) responsive to a plurality of control signals for modifying the spectral content of the input speech signal in each of said selected frequency bands; and

means (27) for combining the modified input speech signal in each of said selected frequency bands to produce said output speech signal.

9. A system in accordance with claim 8, characterised in that said modifying means (30,31) provides a plurality of selectable gains for multiplying the amplitude of the input speech signal by a selected gain factor in each of said selected frequency bands.

10. A system in accordance with any one of claims 2 to 6, characterised in that said modifying means includes

a plurality of second filter means (26) for separating said input speech signal into a plurality of second selected frequency bands;

means (30,31) responsive to a plurality of control signals for modifying the spectral content of the input speech signal in each of said second selected frequency bands; and

means (32) for combining the modified input speech signal in each of said second selected frequency bands to produce said output speech signal.

11. A system in accordance with claim 10, characterised in that said modifying means (26,30,31, 32) provides a plurality of selectable gains for multiplying the amplitude of the input speech signal by a selected gain factor in each of said second selected frequency bands.

12. A system in accordance with claim 6, characterised in that said weighting signal producing means includes

matrix means (28) responsive to said envelope characteristics for multiplying said envelope

characteristics by a plurality of second coefficient values; and

means (29) for combining said multiplied envelope characteristics so as to produce said weighting signals.

13. A system in accordance with claim 12, characterised in that said combining means (Fig. 6) includes

means for combining envelope characteristics multiplied by said first coefficients to produce a plurality of first combined signals;

means for combining said envelope characteristics multiplied by said second coefficients to produce a plurality of second combined signals;

means for determining a plurality of ratios of said plurality of first and second combined signals, said ratios representing said weighting signals.

14. A system in accordance with claim 9, characterised in that said gain factors are selected so as to provide first selected gains when said weighting signals are below selected levels and second selected gains when said weighting signals are at or above said selected levels.

15. A system in accordance with claim 14, characterised in that first selected gains are unity below said selected levels.

16. A system in accordance with claim 15, characterised in that said second selected gains are proportional to W^N, where W is the weighting signal for a selected band and N is a selected exponent.

17. A system in accordance with claim 16, characterised in that N is selected as equal to a value within a range from about 1 to about 3.

18. A system in accordance with claim 17, characterised in that N is selected as equal to 2.

19. A system in accordance with claim 5, characterised in that said envelope detector means (24) detects the peaks of said envelope characteristics and the valleys of said envelope characteristics in each of said frequency bands.

20. A system in accordance with claim 19, characterised by including means (40) for subtracting said valley envelope characteristics from said peak envelope characteristics to form combined envelope characteristics in each said frequency band and said control means in response to said combined envelope characteristics.

21. A method for processing an input speech signal, characterised by the steps of

estimating the spectral shape of said speech signal;

dynamically modifying the spectral shape of said input speech signal in accordance with said estimate to produce an output speech signal.

22. A method in accordance with claim 21, characterised in that said dynamic modification includes the steps of

producing one or more control signals which are functions of said estimate; and

controlling the dynamic modification of the spectral shape of said input speech signal in accordance with said control signals.

23. A method in accordance with claim 22, characterised in that

said estimating steps include the steps of estimating the spectral content of a plurality of first separate frequency bands of said input speech signal relative to the spectral content of one or more of said frequency bands.

24. A method in accordance with claim 23, characterised in that said dynamic modification step includes the step of selecting a filter means having a spectral response specified in accordance with said estimate.

25. A method in accordance with claim 23, characterised in that dynamic modification step includes the step of dynamically modifying the spectral shape of said input speech signal in a plurality of second separate frequency bands in accordance with said estimate.

26. A method in accordance with claim 25, characterised in that the plurality of first separate frequency bands substantially coincides with the plurality of second separate frequency bands.

27. A method in accordance with claim 25, characterised in that the first separate frequency bands are different from the second separate frequency bands.

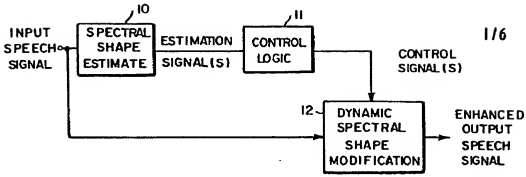

INPUT SPEECH SIGNAL
10
SPECTRAL SHAPE ESTIMATE
ESTIMATION SIGNAL(S)
11
CONTROL LOGIC
CONTROL SIGNAL(S)
12
DYNAMIC SPECTRAL SHAPE MODIFICATION
ENHANCED OUTPUT SPEECH SIGNAL

FIG.1

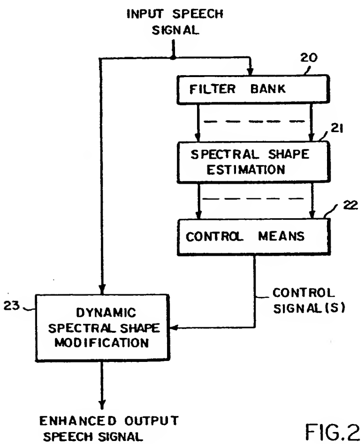
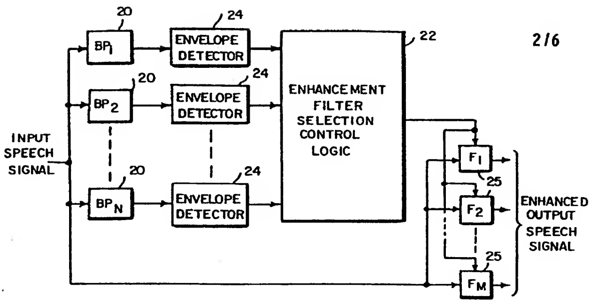
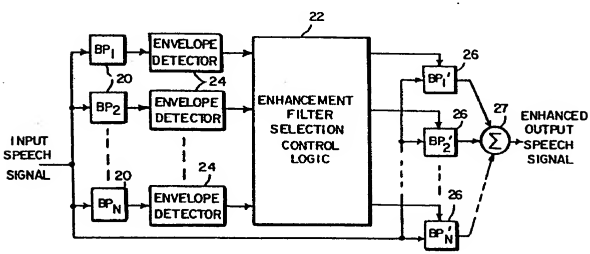

INPUT SPEECH SIGNAL
20
FILTER BANK
21
SPECTRAL SHAPE ESTIMATION
22
CONTROL MEANS
CONTROL SIGNAL(S)
23
DYNAMIC SPECTRAL SHAPE MODIFICATION
ENHANCED OUTPUT SPEECH SIGNAL

FIG.2



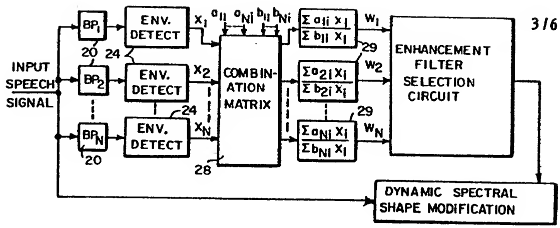
20
24
22
BP1
ENVELOPE DETECTOR
20
24
BP2
ENVELOPE DETECTOR
ENHANCEMENT FILTER SELECTION CONTROL LOGIC
INPUT SPEECH SIGNAL
20
24
BPN
ENVELOPE DETECTOR
F1
25
F2
25
FM
ENHANCED OUTPUT SPEECH SIGNAL

FIG. 3



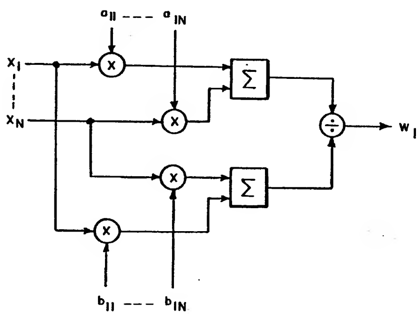
22
BP1
ENVELOPE DETECTOR
20
24
26
BP2
ENVELOPE DETECTOR
ENHANCEMENT FILTER SELECTION CONTROL LOGIC
BP1'
26
27
ENHANCED OUTPUT SPEECH SIGNAL
BP2'
INPUT SPEECH SIGNAL
20
24
26
BPN
ENVELOPE DETECTOR
BPN'

FIG.4



INPUT SPEECH SIGNAL
BP1
BP2
BPN
ENV. DETECT
ENV. DETECT
ENV. DETECT
20
24
X1
X2
XN
a11 aNi b11 bNi
COMBIN-ATION MATRIX
28
Σ a1i x1 / Σ b1i x1
Σ a2i xi / Σ b2i xi
Σ aNi xi / Σ bNi xi
29
W1
W2
WN
ENHANCEMENT FILTER SELECTION CIRCUIT
DYNAMIC SPECTRAL SHAPE MODIFICATION

FIG.5



a11 --- a1N
X1
XN
X
Σ
÷
W1
b11 --- b1N

FIG.6

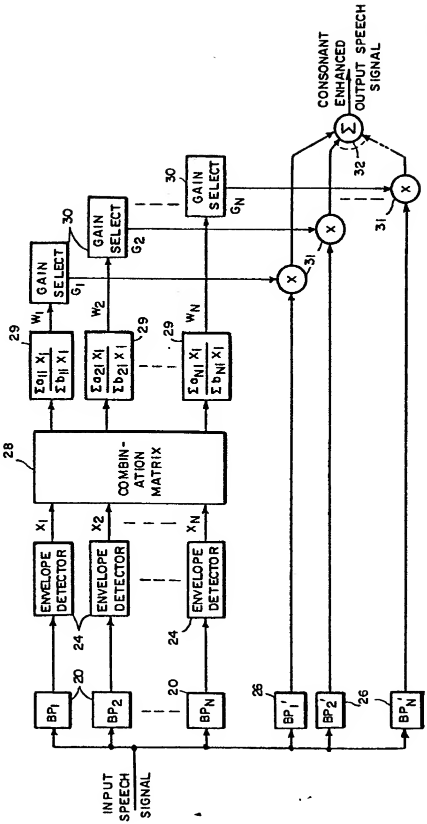

INPUT SPEECH SIGNAL
BP_1
BP_2
BP_N
20
20
BP'_1
BP'_2
BP'_N
26
26
ENVELOPE DETECTOR
ENVELOPE DETECTOR
ENVELOPE DETECTOR
24
24
X_1
X_2
X_N
COMBIN-ATION MATRIX
28
$\frac{\Sigma a_{11} X_i}{\Sigma b_{11} X_i}$
$\frac{\Sigma a_{21} X_i}{\Sigma b_{21} X_i}$
$\frac{\Sigma a_{N1} X_i}{\Sigma b_{N1} X_i}$
29
29
29
W_1
W_2
W_N
GAIN SELECT
GAIN SELECT
GAIN SELECT
30
30
G_1
G_2
G_N
31
31
32
Σ
CONSONANT ENHANCED OUTPUT SPEECH SIGNAL

FIG.7

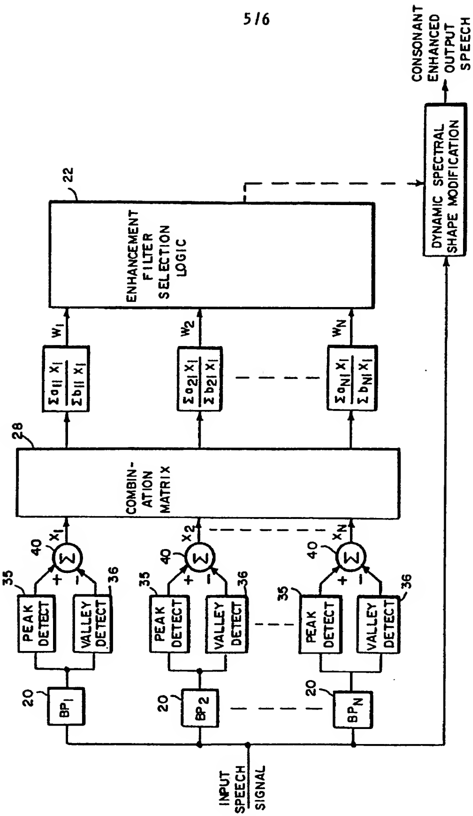

INPUT SPEECH SIGNAL
BP1
BP2
BPN
PEAK DETECT
VALLEY DETECT
X1
X2
XN
COMBIN-ATION MATRIX
Σa11 X1 / Σb11 X1
Σa21 X1 / Σb21 X1
ΣaN1 X1 / ΣbN1 X1
W1
W2
WN
ENHANCEMENT FILTER SELECTION LOGIC
DYNAMIC SPECTRAL SHAPE MODIFICATION
CONSONANT ENHANCED OUTPUT SPEECH SIGNAL
20
35
36
40
28
22

FIG.8

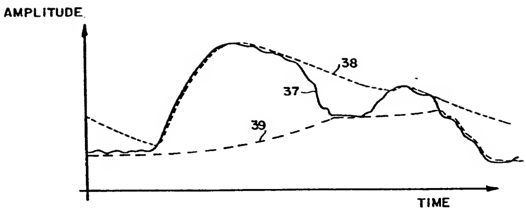


FIG.9

European Patent Office

EUROPEAN SEARCH REPORT

0076687
Application number

EP 82 30 5275

DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int Cl. 3)
X	THE JOURNAL OF THE ACOUSTICAL SOCIETY OF AMERICA, vol.40, no.3, September 1966, New York (US) R.M. GOLDEN: "Improving naturalness and intelligibility of helium-oxygen speech, using vocoder techniques", pages 621-624 * figure 1 *	1-8,10,11,21-25,27	G 10 L 1/00
X	US-A-2 183 248 (R.R. RIESZ) * figure 1 *	1-8,10,11,21-26	
X	DE-A-2 844 979 (J. MANTEL) * figure 4 *	1-6,12,21-23	
			TECHNICAL FIELDS SEARCHED (Int. Cl. 3)
X	DE-A-2 739 609 (L. JONCHERAY et al.) * figure 5 *	1-6,21-23	G 10 L 1/00 H 04 R 25/00
A	FR-A-2 394 865 (A. BARBE) * page 2, line 10 - page 3, line 2 *	1	
A	FR-A-2 226 092 (M. BRUNOT) * claim 1 *	1	

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The present search report has been drawn up for all claims

Place of search	Date of completion of the search	Examiner
THE HAGUE	14-01-1983	ARMSPACH J.F.A.M.

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EUROPEAN SEARCH REPORT

0076687
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DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int. Cl. ³)
A	GB-A-1 384 233 (THE BRITISH BROADCASTING CORP.) * figure 2 * -----	1	

TECHNICAL FIELDS SEARCHED (Int. Cl. ³)

The present search report has been drawn up for all claims

Place of search	Date of completion of the search	Examiner
THE HAGUE	14-01-1983	ARMSPACH J.F.A.M.

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